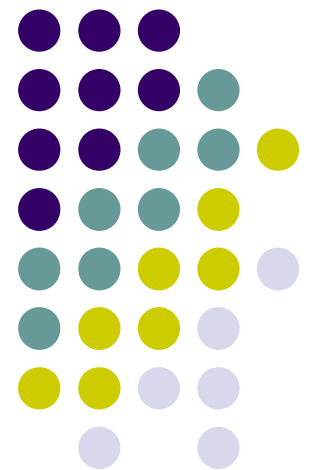


Lossy Medium

EE142

Dr. Ray Kwok

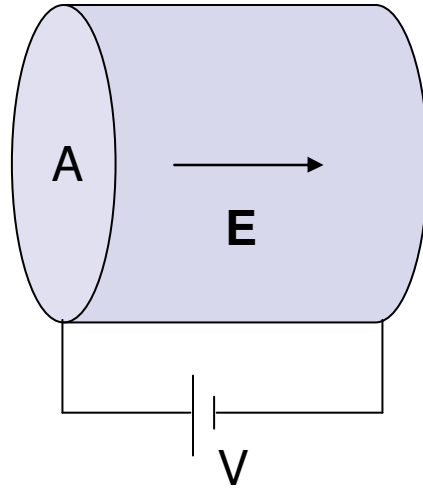


•reference:

Fundamentals of Engineering Electromagnetics, David K. Cheng (Addison-Wesley)
Electromagnetics for Engineers, Fawwaz T. Ulaby (Prentice Hall)



Ohm's Law



$$V = IR$$

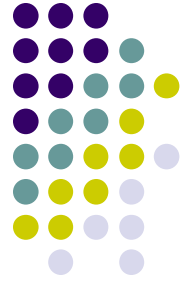
$$El = (JA) \left(\rho \frac{l}{A} \right)$$

$$\vec{E} = \vec{J} \rho \quad \text{resistivity}$$

$$\vec{J} = \frac{1}{\rho} \vec{E}$$

$$\vec{J} = \sigma \vec{E} \quad \text{conductivity}$$

- Low resistivity => “conductor” $\sim < 10^{-5} \Omega \text{ cm}$ (★ T)
- High resistivity => “insulator” $\sim > 10^{10} \Omega \text{ cm}$
- Intermediate resistivity => “semiconductor”
typical $\sim 10^{-3} \text{ to } 10^5 \Omega \text{ cm}$ (★ $e^{E_g/kT}$)
- unit of conductivity = S/m = Siemens/meter = mho/m = $(\Omega \text{ m})^{-1}$



EM Wave through medium

$$\begin{array}{ccc}
 \epsilon \nabla \cdot \vec{E} = \rho_f & & \nabla \cdot \vec{E} = 0 \\
 \nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} & \text{(homogeneous, linear, isotropic)} & \nabla \times \vec{E} = -j\omega\mu\vec{H} \\
 \nabla \cdot \vec{H} = 0 & & \nabla \cdot \vec{H} = 0 \\
 \nabla \times \vec{H} = \vec{J}_f + \epsilon \frac{\partial \vec{E}}{\partial t} & \xrightarrow{\rho \approx 0, J \neq 0} & \nabla \times \vec{H} = \vec{J}_f + j\omega\epsilon\vec{E}
 \end{array}$$

$$\nabla \times \vec{H} = (\sigma + j\omega\epsilon)\vec{E} = j\omega \left(\epsilon + \frac{\sigma}{j\omega} \right) \vec{E} = j\omega\epsilon_c \vec{E}$$

$$\epsilon_c \equiv \epsilon - j\frac{\sigma}{\omega} \equiv \epsilon' - j\epsilon''$$

finite σ means complex ϵ



DIELECTRIC CONSTANTS AND LOSS TANGENTS FOR SOME MATERIALS

Loss Tangent

$$\tan \delta \equiv \frac{\epsilon''}{\epsilon'} = \frac{\sigma}{\omega\epsilon}$$

good conductor $\sigma \gg \omega\epsilon$

good insulator $\sigma \ll \omega\epsilon$

Low $\tan \delta \rightarrow$ low dielectric loss

the smaller, the better !!!

| Material | Frequency | ϵ_r | $\tan \delta$ (25°C) |
|-------------------------|-----------|--------------|----------------------|
| Alumina (99.5%) | 10 GHz | 9.5–10. | 0.0003 |
| Barium tetratitanate | 6 GHz | $37 \pm 5\%$ | 0.0005 |
| Beeswax | 10 GHz | 2.35 | 0.005 |
| Beryllia | 10 GHz | 6.4 | 0.0003 |
| Ceramic (A-35) | 3 GHz | 5.60 | 0.0041 |
| Fused quartz | 10 GHz | 3.78 | 0.0001 |
| Gallium arsenide | 10 GHz | 13. | 0.006 |
| Glass (pyrex) | 3 GHz | 4.82 | 0.0054 |
| Glazed ceramic | 10 GHz | 7.2 | 0.008 |
| Lucite | 10 GHz | 2.56 | 0.005 |
| Nylon (610) | 3 GHz | 2.84 | 0.012 |
| Parafin | 10 GHz | 2.24 | 0.0002 |
| Plexiglass | 3 GHz | 2.60 | 0.0057 |
| Polyethylene | 10 GHz | 2.25 | 0.0004 |
| Polystyrene | 10 GHz | 2.54 | 0.00033 |
| Porcelain (dry process) | 100 MHz | 5.04 | 0.0078 |
| Rexolite (1422) | 3 GHz | 2.54 | 0.00048 |
| Silicon | 10 GHz | 11.9 | 0.004 |
| Styrofoam (103.7) | 3 GHz | 1.03 | 0.0001 |
| Teflon | 10 GHz | 2.08 | 0.0004 |
| Titania (D-100) | 6 GHz | $96 \pm 5\%$ | 0.001 |
| Vaseline | 10 GHz | 2.16 | 0.001 |
| Water (distilled) | 3 GHz | 76.7 | 0.157 |



Example

A sinusoidal E-field with amplitude of 250 V/m and frequency 1 GHz exists in a lossy dielectric medium that has a $\epsilon_r = 2.5$ and loss tangent of 0.001. Find the average power dissipated in the medium per cubic meter.

$$\tan \delta = 0.001 = \frac{\sigma}{\omega \epsilon} = \frac{\sigma}{\omega \epsilon_0 \epsilon_r} = \frac{\sigma}{(2\pi \cdot 10^9) \left(\frac{10^{-9}}{36\pi} \right) (2.5)}$$

$$\sigma = 1.39 \cdot 10^{-4} \quad \text{S/m}$$

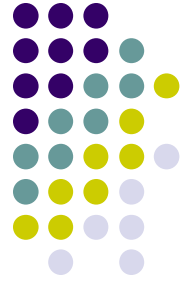
The average power dissipated per unit volume is

$$\frac{P_{\text{ave}}}{V} = \frac{1}{2} \vec{J} \cdot \vec{E} = \frac{1}{2} \sigma E^2 = \frac{1}{2} (1.39 \cdot 10^{-4}) (250)^2$$

$$\frac{P_{\text{ave}}}{V} = 4.34 \quad \text{W/m}^2$$

Note:

$$P_{\text{ave}} = \frac{1}{2} \frac{V^2}{R} = \frac{1}{2} \frac{(El)^2}{\rho l / A} = \frac{1}{2} \sigma E^2 (\ell A)$$



Wave Equation

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\nabla \times \vec{H} = \epsilon_c \frac{\partial \vec{E}}{\partial t}$$

$$\nabla^2 \vec{E} = \mu \epsilon_c \frac{\partial^2 \vec{E}}{\partial t^2}$$

plane wave equation still holds with modification of ϵ

$$\vec{E}(\mathbf{r}, t) = \vec{E}_0 e^{j(\omega t - \mathbf{k}_c \cdot \mathbf{r})} \equiv \vec{E}_0 e^{j\omega t} e^{-\gamma \mathbf{r}}$$

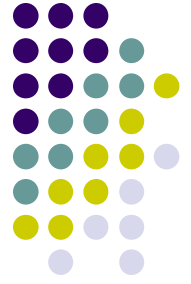
allow k be complex since ϵ is

$$\gamma \equiv jk_c = j\omega \sqrt{\mu \epsilon_c} \equiv \alpha + j\beta$$

propagation constant

attenuation constant

phase constant



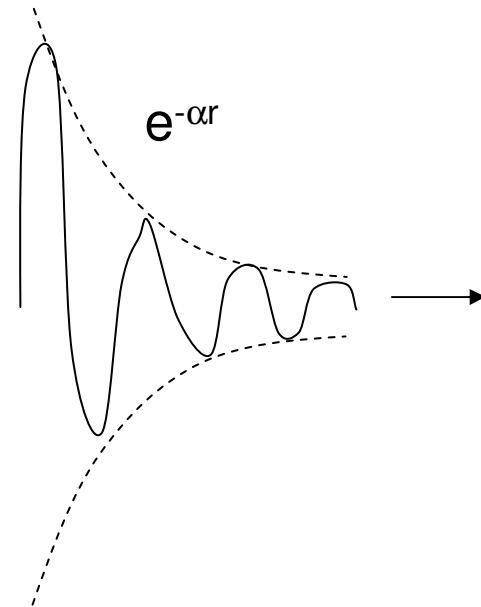
Complex Propagation Constant

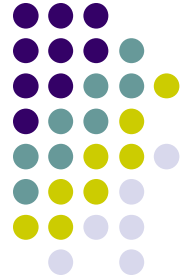
$$\gamma = j\omega\sqrt{\mu\epsilon_c} = j\omega\sqrt{\mu\epsilon\left(1 - j\frac{\epsilon''}{\epsilon'}\right)} = j\omega\sqrt{\mu\epsilon\left(1 - j\frac{\sigma}{\omega\epsilon}\right)} = j\omega\sqrt{\mu\epsilon(1 - j\tan\delta)} = \alpha + j\beta$$

The phasor

$$\vec{E}(\mathbf{r}) = \vec{E}_0 e^{-\gamma r} = \vec{E}_0 e^{-\alpha r} e^{-j\beta r}$$

attenuation





dB scale

power intensity ratio in log scale, not a unit !!

$$(\text{dB}) = 10 \log \left(\frac{I}{I_o} \right) = 10 \log \left(\frac{P}{P_o} \right) = 20 \log \left(\frac{V}{V_o} \right)$$

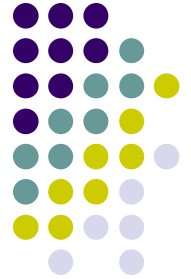
↑
↑
↑
 sound intensity power voltage

> 0 gain
< 0 loss

- 10 log(2) ≈ 3, 3 dB = double
- 10 log(1/2) ≈ -3, -3 dB = half
- 10 log(10) = 10, 10 dB = 10x
- 10 log(100) = 20, 20 dB = 100x
- 10 log(0.1) = -10, -10 dB = 1/10

What is 6 dB? -9 dB? 7 dB? -44 dB?

4x 1/8 5x 4 x 10⁻⁵



dBm & dBW

$$\text{dBW} \equiv 10 \log \left(\frac{P}{1\text{W}} \right)$$

$$\text{dBm} \equiv 10 \log \left(\frac{P}{1\text{mW}} \right)$$

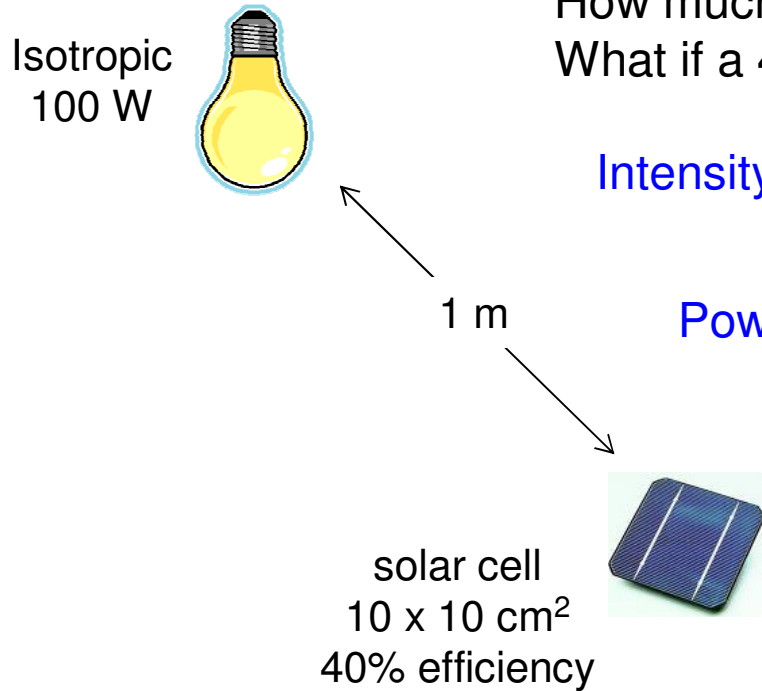
become real units

0 dBm = 1 mW
30 dBW = 1 kW
-30 dBm = 1 μ W

What is 40 dBW? -7 dBm? -26 dBm? 21 dBm?
10 kW 0.2 mW 2.5 μ W 1/8 W



Example



How much electricity generated by the solar cell?
What if a 40 W bulb is used? 200 W bulb?

$$\text{Intensity} = \text{power/area} = \frac{100}{4\pi R^2} = \frac{100}{4\pi(1)^2} = 7.96 \frac{\text{W}}{\text{m}^2}$$

Power generated in solar cell

$$= \left(7.96 \frac{\text{W}}{\text{m}^2}\right) (100\text{cm}^2) (40\%) = 31.8\text{mW}$$

$$\text{In terms of dB} = 10\log\left(\frac{0.0318\text{W}}{100\text{W}}\right) = \boxed{-35\text{dB}}$$

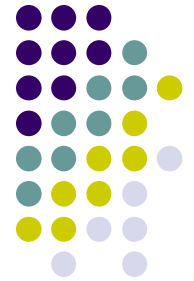
system "gain"

$$40 \text{ W bulb? } -35 = 10\log\left(\frac{P}{40}\right)$$

Power of electricity generated = 12.6 mW

$$200 \text{ W bulb? } -35 = 10\log\left(\frac{P}{200}\right)$$

Power of electricity generated = 63.2 mW



Attenuation

$$\vec{E}(r) = \vec{E}_o e^{-\gamma r} = \vec{E}_o e^{-\alpha r} e^{-j\beta r}$$

$$A(r)[\text{dB}] = 20 \log \left| \frac{\vec{E}(r)}{\vec{E}(0)} \right| = 20 \log |e^{-\alpha r}|$$

$$\log_a(b) = \frac{\log_c(b)}{\log_c(a)}$$

$$A(r)[\text{dB}] = \frac{-20\alpha r}{\ln(10)} = -8.686\alpha r$$

For example, if the electric field intensity going through a medium attenuates at a rate of 0.4 dB/m, what is α ?

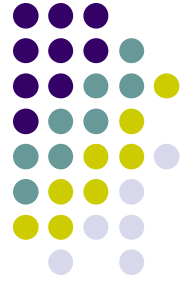
$$-0.4 \text{ dB} = -8.686 \alpha (1 \text{ m})$$

$$\alpha = 0.4/8.686 = 0.046 \text{ nepers/m}$$

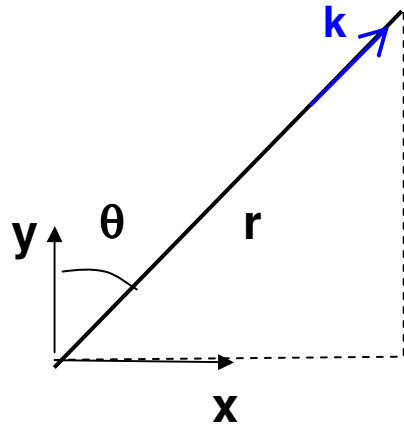
Note: nepers (np) is not a real unit.
similar to radians !!!

Note also α is a positive number for attenuation.

$$\alpha[\text{dB/m}] = 8.686 \alpha[\text{np/m}]$$



Attenuation term



How to express $e^{-\alpha r}$ term??

$$\hat{k} = \hat{x} \sin \theta + \hat{y} \cos \theta$$

$$e^{-\alpha r} = e^{-\alpha(x \sin \theta + y \cos \theta)} = e^{-\alpha \sqrt{x^2 + y^2}}$$

Same ?

$$x \sin \theta + y \cos \theta = x \left(\frac{x}{r} \right) + y \left(\frac{y}{r} \right) = \frac{x^2 + y^2}{\sqrt{x^2 + y^2}} = \sqrt{x^2 + y^2}$$

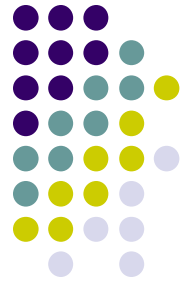
Yes, Q.E.D.

Can think of :

$$\vec{\alpha} \equiv \alpha \hat{k}$$

$$\vec{\alpha} \cdot \vec{r} = \alpha(x \sin \theta + y \cos \theta)$$

$$e^{-\vec{\alpha} \cdot \vec{r}} = e^{-\alpha(x \sin \theta + y \cos \theta)} = e^{-\alpha \sqrt{x^2 + y^2}}$$



Low-loss dielectric ($\epsilon'' \ll \epsilon'$) or ($\sigma \ll \omega\epsilon$)

$$\gamma = j\omega\sqrt{\mu\epsilon\left(1 - j\frac{\epsilon''}{\epsilon'}\right)} = j\omega\sqrt{\mu\epsilon\left(1 - j\frac{\sigma}{\omega\epsilon}\right)} = j\omega\sqrt{\mu\epsilon(1 - j\tan\delta)}$$

$$(1+x)^n = \sum_{k=0}^n \frac{n!}{k!(n-k)!} x^{n-k} = 1 + nx + \frac{n(n-1)}{2} x^2 + \dots$$

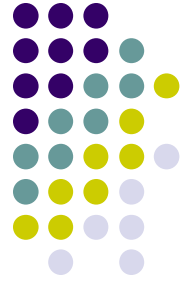
$$(1+x)^n \approx 1 + nx \quad \text{for small } x$$

$$\gamma = j\omega\sqrt{\mu\epsilon}\left(1 - j\frac{\epsilon''}{\epsilon'}\right)^{1/2} \approx j\omega\sqrt{\mu\epsilon}\left(1 - \frac{1}{2}j\frac{\epsilon''}{\epsilon'}\right) \equiv \alpha + j\beta$$

$$\alpha = \frac{\omega\sqrt{\mu\epsilon}}{2} \frac{\epsilon''}{\epsilon'} = \frac{\omega\sqrt{\mu\epsilon}}{2} \frac{\sigma}{\omega\epsilon} = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} \quad \text{small}$$

$$\beta = \omega\sqrt{\mu\epsilon} \approx \omega/v$$

$$\eta_c \equiv \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu}{\epsilon(1 - j\epsilon''/\epsilon')}} = \sqrt{\frac{\mu}{\epsilon}} \left(1 - j\frac{\epsilon''}{\epsilon'}\right)^{-1/2} \approx \sqrt{\frac{\mu}{\epsilon}} \left(1 + \frac{j}{2} \frac{\epsilon''}{\epsilon'}\right) \approx \sqrt{\frac{\mu}{\epsilon}}$$



Good Conductor ($\sigma \gg \omega\epsilon$)

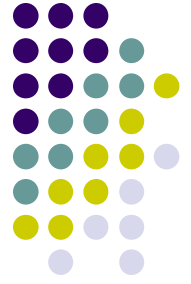
$$\gamma = j\omega \sqrt{\mu\epsilon \left(1 - j \frac{\sigma}{\omega\epsilon}\right)} \approx j\omega \sqrt{-j \frac{\sigma}{\omega\epsilon} \mu\epsilon}$$

$$\sqrt{-j} = \sqrt{e^{-j\pi/2}} = e^{-j\pi/4} = \frac{1-j}{\sqrt{2}}$$

$$\gamma = j\omega \frac{1-j}{\sqrt{2}} \sqrt{\frac{\mu\sigma}{\omega}} = (j+1) \sqrt{\frac{\mu\sigma\omega}{2}} \equiv \alpha + j\beta$$

$$\alpha = \beta = \sqrt{\frac{\mu\sigma\omega}{2}}$$

$$\eta_c \equiv \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu}{\epsilon(1 - j\sigma/\omega\epsilon)}} \approx \sqrt{j \frac{\mu\omega\epsilon}{\epsilon\sigma}} \approx \frac{1+j}{\sqrt{2}} \sqrt{\frac{\mu\omega}{\sigma}} = (1+j) \sqrt{\frac{\mu\omega}{2\sigma}} = \boxed{(1+j) \frac{\alpha}{\sigma}}$$



Skin Depth δ

$$\delta \equiv \frac{1}{\alpha} = \sqrt{\frac{2}{\mu\sigma\omega}} = \delta_s$$

(NOT loss tangent δ !!!!!)

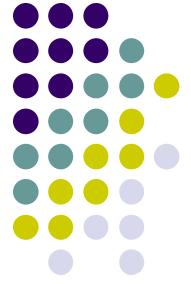
$$\vec{E}(\mathbf{r}) = \vec{E}_0 e^{-\alpha r} e^{-j\beta r}$$

At $r = \delta$, $|E|$ decreases to $1/e$ (or 63% drop).

$$A(r)[\text{dB}] = 20 \log \left| \frac{\vec{E}(r)}{\vec{E}(0)} \right| = 20 \log |e^{-\alpha r}| = -8.686 \alpha r$$

At $r = \delta$, $|E|$ decreases by -8.7 dB.

At $r = 2\delta$, $|E|$ decreases by -17.3 dB....



General Material

$$\gamma = j\omega\sqrt{\mu\epsilon_c} = j\omega\sqrt{\mu(\epsilon' - j\epsilon'')} \equiv \alpha + j\beta$$

$$\gamma^2 = -\omega^2\mu(\epsilon' - j\epsilon'') = \alpha^2 - \beta^2 + 2j\alpha\beta$$

$$-\omega^2\mu\epsilon' = \alpha^2 - \beta^2 \quad \text{real}$$

$$j\omega^2\mu\epsilon'' = j2\alpha\beta \quad \text{imaginary}$$

$$\alpha^2 = \beta^2 - \omega^2\mu\epsilon' = \left(\frac{\omega^2\mu\epsilon''}{2\alpha}\right)^2 - \omega^2\mu\epsilon'$$

$$4\alpha^4 + 4\alpha^2\omega^2\mu\epsilon' - (\omega^2\mu\epsilon'')^2 = 0$$

$$\alpha^2 = \frac{-4\omega^2\mu\epsilon' \pm \sqrt{(4\omega^2\mu\epsilon')^2 + 16(\omega^2\mu\epsilon'')^2}}{8}$$

$$\alpha^2 = \frac{\omega^2\mu\epsilon'}{2} \left(-1 \pm \sqrt{1 + \tan^2 \delta}\right)$$

$$\alpha^2 = \frac{\omega^2\mu\epsilon'}{2} \left(\sqrt{1 + \tan^2 \delta} - 1\right)$$

$$\beta^2 = \left(\frac{\omega^2\mu\epsilon''}{2\alpha}\right)^2 = \frac{\omega^2\mu\epsilon' \tan^2 \delta}{2(\sqrt{1 + \tan^2 \delta} - 1)}$$

$$\beta^2 = \frac{\omega^2\mu\epsilon' \tan^2 \delta}{2(\sqrt{1 + \tan^2 \delta} - 1)(\sqrt{1 + \tan^2 \delta} + 1)}$$

$$\beta^2 = \frac{\omega^2\mu\epsilon' \tan^2 \delta (\sqrt{1 + \tan^2 \delta} + 1)}{2(1 + \tan^2 \delta - 1)}$$

$$\beta^2 = \frac{\omega^2\mu\epsilon'}{2} (\sqrt{1 + \tan^2 \delta} + 1)$$

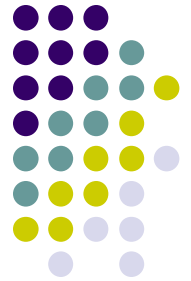
$$\eta_c = \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu}{\epsilon'(1 - j \tan \delta)}}$$

$$\eta_c = \sqrt{\frac{\mu}{\epsilon'}} (1 - j \tan \delta)^{-1/2}$$



Summary

| | Any Medium | Lossless Medium ($\sigma = 0$) | Low-loss Medium ($\varepsilon''/\varepsilon' \ll 1$) | Good Conductor ($\varepsilon''/\varepsilon' \gg 1$) | Units |
|---|---|-------------------------------------|---|--|--------------|
| $\alpha =$ | $\omega \left[\frac{\mu\varepsilon'}{2} \left[\sqrt{1 + \left(\frac{\varepsilon''}{\varepsilon'}\right)^2} - 1 \right] \right]^{1/2}$ | 0 | $\frac{\sigma}{2} \sqrt{\frac{\mu}{\varepsilon}}$ | $\sqrt{\pi f \mu \sigma}$ | (Np/m) |
| $\beta =$ | $\omega \left[\frac{\mu\varepsilon'}{2} \left[\sqrt{1 + \left(\frac{\varepsilon''}{\varepsilon'}\right)^2} + 1 \right] \right]^{1/2}$ | $\omega\sqrt{\mu\varepsilon}$ | $\omega\sqrt{\mu\varepsilon}$ | $\sqrt{\pi f \mu \sigma}$ | (rad/m) |
| $\eta_c =$ | $\sqrt{\frac{\mu}{\varepsilon'}} \left(1 - j\frac{\varepsilon''}{\varepsilon'}\right)^{-1/2}$ | $\sqrt{\frac{\mu}{\varepsilon}}$ | $\sqrt{\frac{\mu}{\varepsilon}}$ | $(1 + j)\frac{\alpha}{\sigma}$ | (Ω) |
| $u_p =$ | ω/β | $1/\sqrt{\mu\varepsilon}$ | $1/\sqrt{\mu\varepsilon}$ | $\sqrt{4\pi f/\mu\sigma}$ | (m/s) |
| $\lambda =$ | $2\pi/\beta = u_p/f$ | u_p/f | u_p/f | u_p/f | (m) |
| Notes: $\varepsilon' = \varepsilon$; $\varepsilon'' = \sigma/\omega$; in free space, $\varepsilon = \varepsilon_0$, $\mu = \mu_0$; in practice, a material is considered a low-loss medium if $\varepsilon''/\varepsilon' = \sigma/\omega\varepsilon < 0.01$ and a good conducting medium if $\varepsilon''/\varepsilon' > 100$. | | | | | |



Example - The skin depth of a certain nonmagnetic conducting material is $2\mu\text{m}$ at 5 GHz. Determine the phase velocity in the material. What is the attenuation (in dB) when the wave penetrates $10\mu\text{m}$ into the material?

$$\text{phase velocity } v = \omega/\beta$$

$$\text{for conductor, } \alpha = \beta = 1/\delta$$

$$v = \omega\delta = (2\pi)(5 \times 10^9) (2 \times 10^{-6}) = 6.28 \times 10^4 \text{ m/s}$$

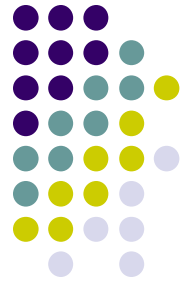
$$A(r)[\text{dB}] = 20 \log \left| \frac{\vec{E}(r)}{\vec{E}(0)} \right| = 20 \log |e^{-\alpha r}| = -8.686\alpha r$$

$$A(r)[\text{dB}] = -8.686r / \delta = -8.686(10 / 2) = -43.4\text{dB}$$

in just 5 skin depth.

(- 43 dB = 1 / 20,000 !!!)

Only surface current on conductors.



Example — (a) Calculate the dielectric loss (in dB) of an EM wave propagating through 100 m of teflon at 1 MHz. (b) at 10 GHz ?

Teflon: $\epsilon_r = 2.08$, $\tan\delta = 0.0004$ at 25°C assuming frequency independence.

$$(a) \quad \tan \delta = \frac{\sigma}{\omega\epsilon}$$

$$\sigma = \omega\epsilon_0\epsilon_r \tan \delta = (2\pi \cdot 10^6) \left(\frac{10^{-9}}{36\pi} \right) (2.08)(0.0004) = 4.6 \cdot 10^{-8} \quad \text{S/m}$$

$$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} = \frac{\sigma\eta_0}{2\sqrt{\epsilon_r}} = \frac{(4.6 \cdot 10^{-8})(377)}{2\sqrt{2.08}} = 6.04 \cdot 10^{-6} \quad \text{np/m}$$

$$A(\text{dB}) = -8.686\alpha r = -8.686(6.04 \cdot 10^{-6})(100) = -0.005 \quad \text{dB}$$

$$(b) \quad \sigma = \omega\epsilon_0\epsilon_r \tan \delta = (2\pi \cdot 10^{10}) \left(\frac{10^{-9}}{36\pi} \right) (2.08)(0.0004) = 4.6 \cdot 10^{-4} \quad \text{S/m}$$

$$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} = \frac{\sigma\eta_0}{2\sqrt{\epsilon_r}} = \frac{(4.6 \cdot 10^{-4})(377)}{2\sqrt{2.08}} = 6.04 \cdot 10^{-2} \quad \text{np/m}$$

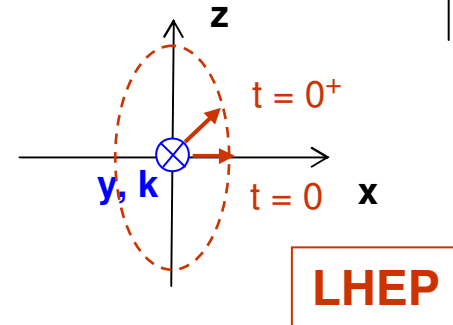
$$A(\text{dB}) = -8.686\alpha r = -8.686(6.04 \cdot 10^{-2})(100) = -50 \quad \text{dB}$$

Coaxial cable works well at low freq (TV to antenna) but not so well at high freq. !!

Example — In a nonmagnetic, lossy, dielectric medium, a 300-MHz plane wave is characterized by the magnetic field phasor $\vec{H} = (\hat{x} - j4\hat{z})e^{-2y}e^{-j9y}$ A/m. Obtain time-domain expressions for the electric and magnetic field vectors. What is the polarization state of this wave?

$$\vec{H}(\vec{r}, t) = \Re\{(\hat{x} - j4\hat{z})e^{-2y}e^{j(\omega t - 9y)}\}$$

$$\vec{H}(\vec{r}, t) = \hat{x}e^{-2y} \cos(\omega t - 9y) + \hat{z}4e^{-2y} \sin(\omega t - 9y)$$



$$\alpha = 2, \quad \beta = 9$$

$$-\omega^2\mu\epsilon' = \alpha^2 - \beta^2$$

$$\omega^2\mu\epsilon'' = 2\alpha\beta$$

$$\frac{\epsilon''}{\epsilon'} = \frac{2\alpha\beta}{\beta^2 - \alpha^2} = \frac{2(2)(9)}{9^2 - 2^2} = 0.468 = \tan \delta$$

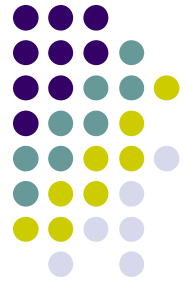
$$\epsilon_r = \frac{\epsilon'}{\epsilon_0} = \frac{\beta^2 - \alpha^2}{\omega^2\mu_0\epsilon_0} = \frac{77c^2}{\omega^2} = \frac{77(3 \cdot 10^8)^2}{(2\pi \cdot 300 \cdot 10^6)^2} = 1.95$$

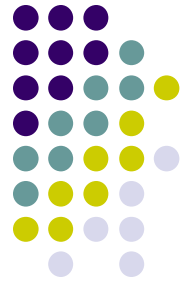
$$\eta_c = \sqrt{\frac{\mu}{\epsilon'}}(1 - j \tan \delta)^{-1/2} = \frac{\eta_0}{\sqrt{\epsilon_r}}(1 - j \tan \delta)^{-1/2} = \frac{377}{\sqrt{1.95}}(1 - j0.468)^{-1/2}$$

$$\eta_c = 257 \angle 12.5^\circ = 257e^{j0.22}$$

$$\vec{E} = \hat{z}\eta_c e^{-2y} \cos(\omega t - 9y) - \hat{x}4\eta_c e^{-2y} \sin(\omega t - 9y)$$

$$\vec{E} = \hat{z}257e^{-2y} \cos(\omega t - 9y + 0.22) - \hat{x}1028e^{-2y} \sin(\omega t - 9y + 0.22)$$





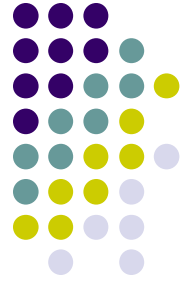
Exercise (how to write the attenuation?)

Given E_o at the origin has a amplitude of 1 V/m along the y-axis in a non-magnetic medium, with the propagation given by:

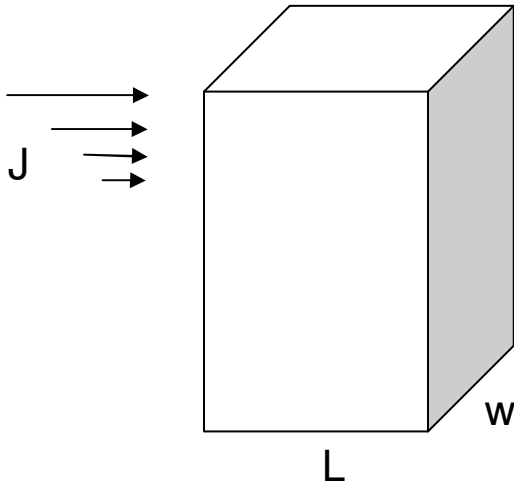
$$\sim \sin(\omega t + 30x - 15z)$$

$$\epsilon = (4 - j0.02)\epsilon_o$$

Write $\vec{H}(\vec{r}, t) = ?$

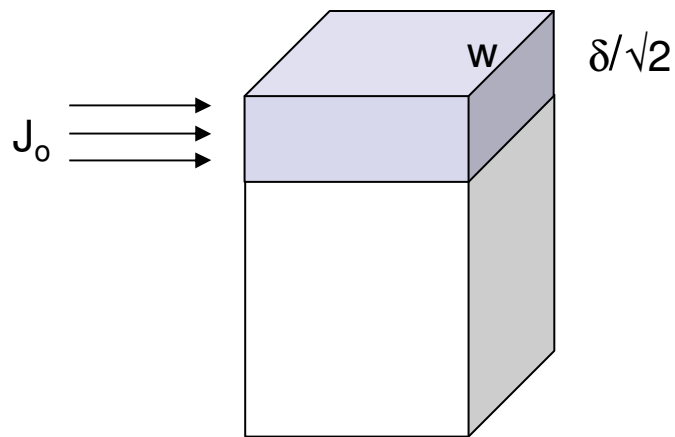


Surface resistance for conductors



$$J = \sigma E = J_0 e^{-\alpha z} e^{-j\beta z} = J_0 e^{-(1+j)z/\delta}$$

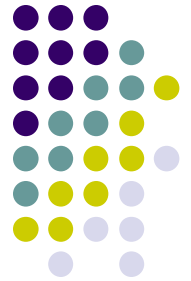
$$I = \int \vec{J} \cdot d\vec{a} = J_0 w \int_0^\infty e^{-(1+j)z/\delta} dz = \frac{J_0 w \delta}{(1+j)} = \frac{J_0 w \delta}{\sqrt{2}} e^{-j\pi/4}$$



$$V = E_0 L = \frac{J_0}{\sigma} L \quad \text{on surface}$$

$$Z = \frac{V}{I} = \frac{1+j}{\sigma \delta} \frac{L}{w} \equiv Z_s \frac{L}{w} \quad \text{similar to } R \text{ \& } \rho$$

$$Z_s = \frac{1+j}{\sigma \delta} \quad \text{surface impedance } (\Omega)$$



Homework

1. Determine the frequency at which a time-harmonic electric field intensity causes a conduction current density and a displacement current density of equal magnitude in
 - (a) seawater with $\epsilon_r = 72$ and $\sigma = 4$ S/m, and
 - (b) moist soil with $\epsilon_r = 2.5$ and $\sigma = 10^{-3}$ S/m.

2. Calculations concerning the electromagnetic effect of currents in a good conductor usually neglect the displacement current even at microwave frequencies.
 - (a) Assuming $\epsilon_r = 1$ and $\sigma = 5.7 \times 10^7$ S/m for copper, compare the magnitude of the displacement current density with that of the conduction current density at 100 GHz.
 - (b) Write the differential equation in phasor form for magnetic field intensity H in a source-free good conductor.